

Assignment

To design a 16-bit ALU which will be used in the datapath of the microprocessor. This ALU must support two's complement arithmetic and the instructions in the baseline architecture.

Description

The Arithmetic and Logic Unit (ALU) is the heart of your processor. The minimum set of instructions your ALU needs to support consists of

(i) ADD, (ii) SUB, (iii)CMP, (iv) AND, (v) OR, and (vi) XOR.

Your ALU must support both register-based operands and immediate operands.

You can implement all the above instructions with the adder as the basic building block. A two's complement subtractor and a comparator can be derived from the adder structure. Two's complement subtraction ($A - B$) is implemented by adding A , B_{bar} , and 1 ($\text{Carry}_{\text{in}} = 1$). Compare functions (GE, LE), which are used to set processor flags for conditional branches, can be implemented with a subtractor and the information from the most significant bit (sign bit). Implementing "Less Than" is a little more complicated because of overflow problems. To detect a zero output, you need a NOR tree to verify that no result bit is a 1.

There are two general approaches in designing the ALU. The first one is conceptually simpler but the second one may result in a more efficient implementation.

1. Separate Logic Blocks

Your ALU must perform six different functions; this could be accomplished as shown below, with the adder unit performing three of these, and simple logic blocks performing the other three. Then a four-to-one multiplexer (or tristate buffers) could be used to select the correct function.

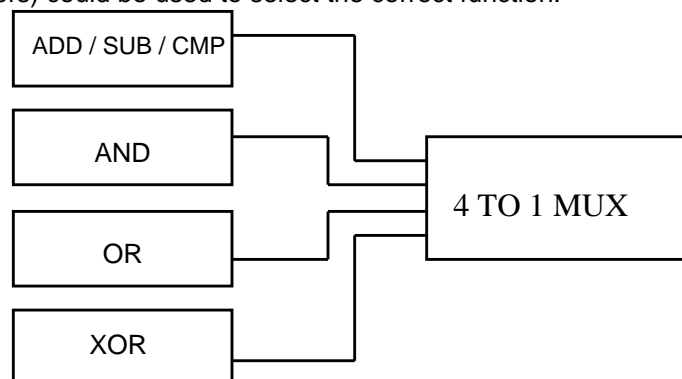


Figure 1: Simple ALU approach.

2. Modified Ripple Carry Adder

An alternative is to make the adder perform the logic functions as well as arithmetic functions. Consider the standard adder equations:

$$\text{Sum: } S = ABC + AB'C' + A'B'C + A'BC'$$

$$\text{Carry: } \text{Cout} = AB + BC + AC$$

If we define

$$\text{Half sum: } H = AB' + A'B$$

then we could write the adder equations as

$$\text{Sum: } S = HC' + H'C$$

$$\text{Carry: } \text{Cout} = AB + HC$$

From the equations it is clear that when C is held at logical 0, the sum output is an XOR of A and B .

$$S = AB' + A'B \text{ (XOR operation)}$$

Also, when C is held at logical 0, then,

$$\text{Cout} = AB \text{ (AND operation)}$$

When C is logic 1, then

$$C_{out} = AB + A + B = A + B \text{ (OR operation)}$$

This shows that appropriate switching of the carry line between adder elements will give the ALU logical functions. An example block diagram is shown below.

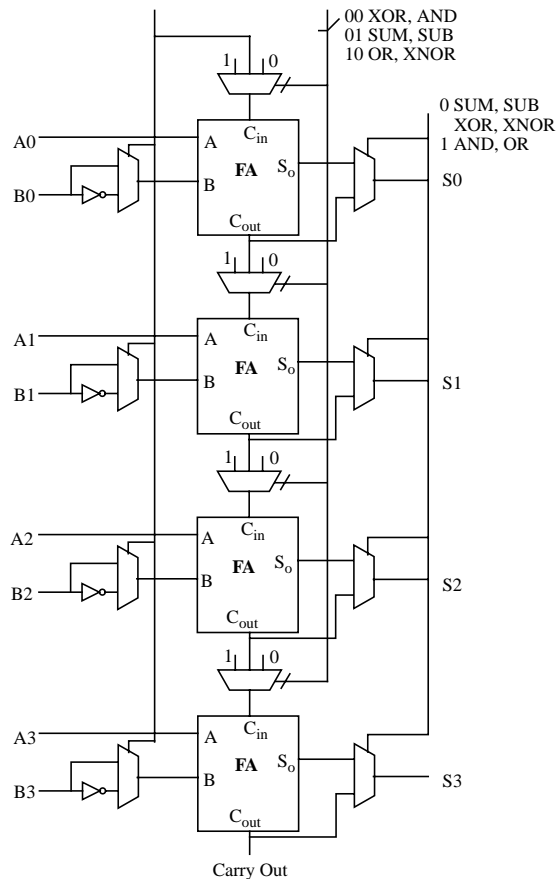


Figure 2: Merged arithmetic/Logic functions.

3. Adder Designs

Any of a variety of 16-bit adder designs could be used in your processors. Whichever design you choose to implement, you would first build a 1-bit cell. Note that any number of such elements may be cascaded to form an adder of desired width.

The truth table of a 1-bit adder is as follows:

C	A	B	Cout	S
0	0	0	0	0
0	0	1	0	1
0	1	0	0	1
0	1	1	1	0
1	0	0	0	1
1	0	1	1	0
1	1	0	1	0
1	1	1	1	1

Hence, we may write the standard adder equation in the form

$$\text{Sum} \quad S = ABC + AB'C' + A'B'C + A'BC'$$

$$\text{Carry} \quad C_{out} = AB + BC + AC = AB + C(A+B)$$

We now discuss two different kinds of full adders:

1. Transmission Gate Adder

Fig. 3 is the schematic of a transmission gate adder. The XOR function is implemented on the left. Its outputs are then used to implement the sum and the carry. It has 4 transmission gates, 4 inverters and an XOR gate. Its advantages include equal SUM and CARRY delay times, and non-inverted SUM and CARRY output signals.

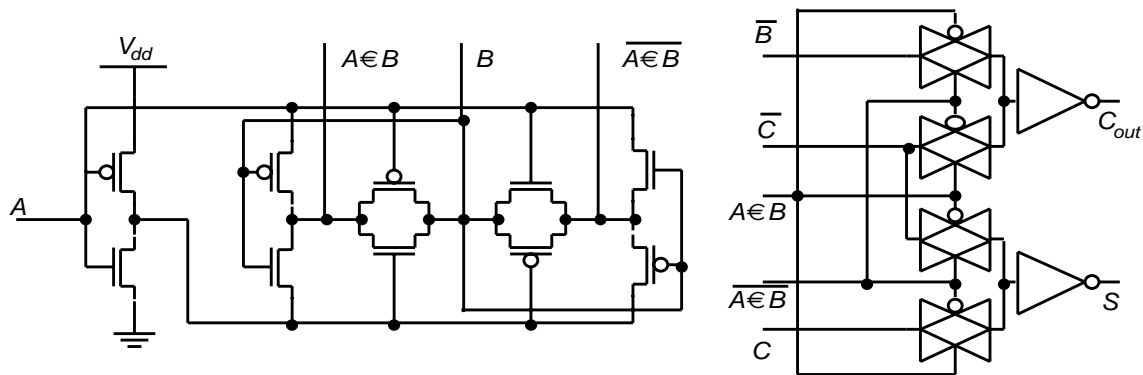


Figure 3: Transmission Gate Adder.

3. Combinational Adder

One example of a combinational adder is shown in Fig. 4. Of course, there are many ways to implement the adder equations, depending upon which gates one chooses to use. In this example, the CARRY signal is used in the generation of SUM (so SUM will be delayed with respect to CARRY). It has 24 transistors, the same as the transmission gate adder.

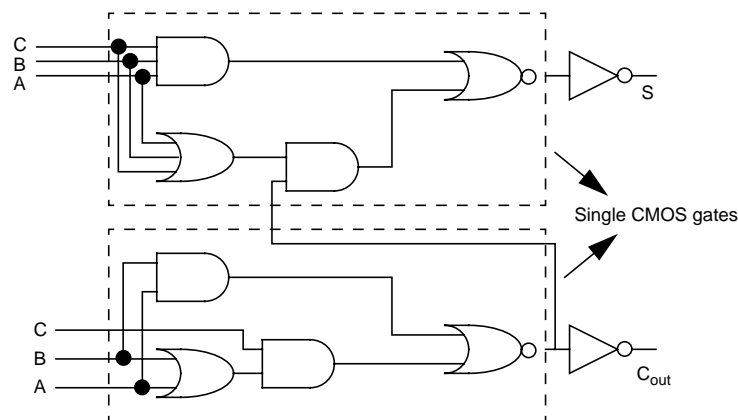


Figure 4: Combinational Adder.

Procedure

Your ALU must be bitslice width matched to your register file. In this assignment, you are not required to implement the logic to set processor flags (zero, overflow, carry, etc.), but you should have in mind the requirements so that you do not have to rework your design later (e.g., how will you detect an overflow?).

1. Schematic Design

The ALU is the key block of your datapath and can be the majority of delay in the execute cycle. This hand-out assumes a ripple-carry adder based ALU with a few different choices for the full-adder design. You may also choose to implement any of the carry lookahead schemes discussed in lecture (basic CLA, carry-select, log lookahead/Brent-Kung). These will be less regular and will require significantly more layout effort. Once you choose a particular adder, optimize it for speed. Turning in a design with all minimum-sized devices amounts to a simple logic switch design and will result in significant deductions.

You can implement a normal ripple-carry adder with complex gates, as shown in Fig. 5. Since the critical

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CAD 4

2. Logic Verification

3. Layout

Be sure to run the buses over your cells. Bitslice width matching with other datapath modules is very important. Save vertical routing resources for datapath integration. In the layout of a fast adder, you may find this particularly difficult to do. If you find it necessary to use a lot of vertical routing resources within the leaf cell, you could opt to place the ALU on the top or bottom of the datapath such that fewer signals must run over the ALU.

4. Design Verification

5. Analog Simulation

6. Add Delays

7. Final Digital Simulation

Re-simulate the logic in Modelsim to verify that it works satisfactorily after the delays have been added.

Comments

- You can implement the ALU any way you wish. Remember that it is a datapath element. You can choose any select-line set-up that you wish (i.e. you are not restricted to any type of encoding). Modelsim traces should show that the 16-bit ALU works for all six (more if you implement more) functions with a few tests per function. It's fine to implement carry-lookahead schemes but note that they will be less regular and will require more layout work. Again, don't use any dynamic circuits (e.g. no dynamic Manchester carry chains).
- You are allowed to add instructions to the basic instruction set, but do not remove any.
- Keep in mind that you are going to need to implement some Processor Status Register (PSR) bits. You implement the PSR later in the semester. However, some of the groups might wish to place the PSR on the datapath. If this is the case, you might want to implement the PSR in this CAD assignment since you are familiar with the ALU and time will be tight later in the semester. See the **PSR Description** provided at the end of this assignment for more information.

Requirements

All files should be placed in your group directory in the directory **cad4**.

- 16-bit ALU (alu16) schematic and layout plots. The schematic should show delays added on to the delay element. You should also have the schematic and layout for the 1-bit ALU (alu1).
- Modelsim .wlf and .do for the 16-bit ALU with delays. These should demonstrate all ALU functions.
- Eldo config file(s) and wave.jpgs showing the longest rise and fall delays for each output of the 1-bit ALU or full 16-bit ALU. 1-bit ALU Eldo is fine for ripple carry implementations but you must do 16-bit for anything else. Explain the delays; i.e., which path was critical in determining the delays.
- DRC report from ICrules.
- LVS report from ICtrace(M).
- PEX reports from ICextract(M).
- A README explaining the choice of the adder style and floor planning, along with the names/descriptions of the various files you're submitting. Feel free to include any other pertinent comments (e.g., placement of PSR in the control or the datapath).

Appendix: PSR Description

The Processor Status Register is a 16 bit scannable register that holds information pertaining to the current state of the microprocessor. The assignments of bits in the PSR is shown below.

15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
R	R	R	R	I	P	E	0	N	Z	F	0	0	L	T	C

Bits 12 to 15 are "Reserved" which means that these bits should not be written or read. Bits 3, 4 and 8 may be hardwired to zero. The bits **N**egative, **L**ow, **Z**ero, **C**arry and **O**ver**F**low are set by the various arithmetic instructions. Unless you are implementing maskable interrupts, you can set the **E** and **I** bits to zero. Similarly, the **T**race and **P** bits can be set to zero unless you implement tracing. See the notes at the end of the handout on the instruction set for more information about these flags

Adding or subtracting two n-bit numbers can require an n+1 bit number to fully express the result. When the result requires more bits than are available, we have an overflow or carry condition. Overflow occurs for

two's complement numbers under the conditions indicated in Table 1. An easy way to determine overflow

Table 2: Overflow Conditions.

Operation	Operand A	Operand B	Result
A + B	≥ 0	≥ 0	< 0
A + B	< 0	< 0	≥ 0
A - B	≥ 0	< 0	< 0
A - B	< 0	≥ 0	≥ 0

is to exclusive-OR the carry-in of the high-order bit with the carry-out of the high-order bit. It is left as an exercise for the student to verify that this detects overflow.

Carry occurs when the result of adding unsigned n-bit numbers exceeds 2^n-1 , or the result of subtracting unsigned numbers is negative.

The only way to determine Zero is to check that all of the bits are zero with an OR or OR-like function. The Negative bit can be set from the high-order bit. The L bit is set assuming unsigned numbers, i.e., that all 16 bits of the operand indicate values, rather than sign, or, that all numbers are positive. The compare instruction does a subtraction of the value in *src1* register from that in the *src2* register. It is not possible to get an overflow when numbers of the same sign are subtracted. In the compare instruction, operands are treated as both signed and unsigned integers at the same time; the processor does not know which they are. The programmer does know what type of data the operands represent, so can choose the correct condition on which to conditionally branch or jump.

Be sure to clear the flags when they should be cleared (as well as setting them at the appropriate times). Do not change flags when they should not be changed. This can be implemented by clocking the flip-flop for a particular flag bit whenever an instruction which is to set or clear the bit has been executed.

You could include instructions for loading and storing the processor status register (LPR and SPR). These instructions would move data from one of the general-purpose registers into the status register, or read the status register into a general-purpose register. These instructions are not included in the baseline machine; the baseline machine requires the PSR to be modified only by arithmetic instructions and reset. If you are not implementing the LPR and SPR instructions, you can save area by just implementing the flag flip-flops which you need (5 for the baseline processor).

Implementation of Flags

The flags for addition are straightforward. Let c_n be the carry out of the most significant bit and c_{n-1} be the carry into the most significant position. Then $C = c_n$, and $F = c_n \oplus c_{n-1}$. The most common way of doing subtraction when two's complement numbers are used is to add the one's complement of the number to be subtracted and make the carry in to the least significant bit equal to one. In this case the carry flag is given by $C = \overline{c_n}$, and F is the same as for addition. In the compare operation (CMP, CMPI) $L = \overline{c_n}$, and N can be derived in several ways:

$$N = a_{n-1} \oplus b_{n-1} \oplus \overline{c_n} = a_{n-1} \oplus \overline{b_{n-1}} \oplus c_n = s_{n-1} \oplus c_n \oplus c_{n-1}$$

where $n-1$ refers to the most significant bits in computing A-B, and s_{n-1} is the sum bit.